Saturated Output Regulation of Distributed Parameter Systems with Collocated Actuators and Sensors

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Outline







 Main Problem
 Systems with Input Saturation

 Results
 Reference and disturbance signals

 Example
 Semi-global output regulation problem







Main Problem	
Results	Reference and disturbance signals
Example	

- In reality, it is natural to consider limitations in the actuators.
- Linear systems subject to input saturation

$$\dot{x}(t) = Ax(t) + B\phi(u(t)) + B_d w_d(t), \quad x(0) = x_0,$$

 $y(t) = Cx(t).$

Here ϕ is a saturation function where the input u(t) takes values in an interval $[u_{min}, u_{max}]$.

• Output regulation of infinite-dimensional linear systems subject to input saturation:

For given $y_{ref}(t)$ and $w_d(t)$, construct u(t) so that

$$\|y(t) - y_{\mathrm{ref}}(t)\| \to 0, \qquad \text{as} \quad t \to \infty$$

for all $x_0 \in X$.

Main Problem	Systems with Input Saturation
Results	Reference and disturbance signals
Example	

We consider class of abstract SISO systems of the form

$$\dot{x}(t) = Ax(t) + B\phi(u(t)) + B_d w_d(t), \quad x(0) = x_0,$$

 $y(t) = B^*x(t)$

on a real Hilbert space X. Here $x(t) \in X$ is the state, $u(t) \in \mathbb{R}$ the input, $y(t) \in \mathbb{R}$ the output, and $w_d(t) \in \mathbb{R}^{n_d}$ an external disturbance.

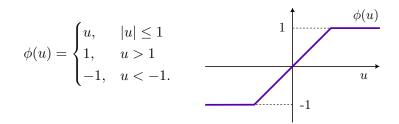
Main Problem	Systems with Input Saturation
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Example	Semi-global output regulation problem

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Comments

- The considered class of systems arise in the study of systems with collocated actuators and sensors.
- Output regulation has been studied actively for finite-dimensional saturated systems. For DPS, results mainly focus on PI control (constant y_{ref}(t) and w_d(t)), esp. [Logemann-Ryan-Townley '98-, Logemann et. al., '00 Oostveen '00]
- We consider time-varying $y_{ref}(t)$ and $w_d(t)$. Our results extend the finite-dimensional theory in [Saberi et al., 2003].

The Reference and Disturbance Signals

We consider references $y_{\mathrm{ref}}(t)$ and disturbances $w_d(t)$ of the forms

$$y_{\rm ref}(t) = \sum_{k=1}^{q_0} a_k \cos(\omega_k + \theta_k), \qquad w_d(t) = \sum_{k=1}^{q_0} b_k \cos(\omega_k + \varphi_k)$$

for **known** frequencies ω_k , amplitudes a_k , b_k , and phases θ_k , φ_k . Such $y_{ref}(t)$ and $w_d(t)$ can be generated by **an exosystem**

$$\dot{v}(t) = Sv(t), \quad v(0) = v_0 \in \mathbb{R}^q,$$

$$w_d(t) = Ev(t),$$

$$y_{ref}(t) = -Fv(t)$$

$$S, E, F$$

where $S \in \mathbb{R}^{q \times q}$, $F \in \mathbb{R}^{1 \times q}$, $E \in \mathbb{R}^{n_d \times q}$, and v_0 are determined by ω_k , a_k , b_k , θ_k , and φ_k . In particular, $\sigma(S) = \{\pm i\omega_k\}_{k=1}^q$.

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Semi-Global Output Regulation Problem.

Consider a compact set $\mathcal{W}_0 \subset \mathbb{R}^q$. Find a linear control law in the form

$$u(t) = -\kappa y(t) + Lv(t)$$

such that $\kappa > 0, \ L \in \mathbb{R}^{1 \times q}$ and

 $\textbf{0} \quad 0 \in X \text{ is a globally asymptotically stable equilibrium of system}$

$$\dot{x}(t) = Ax(t) + B\phi(-\kappa y(t)), \qquad x(0) = x_0.$$

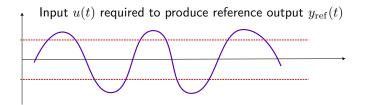
② For all $x_0 \in X$ and $v_0 \in W_0$, the error between the output y(t) and the reference signal $y_{ref}(t)$ satisfies

$$\lim_{t \to \infty} |y(t) - y_{ref}(t)| = 0.$$

Comment: u(t) is based on output feedback + $y_{ref}(t)$ and $w_d(t)$.

Why "Semi-global"?

"For all $x_0 \in X$ and $v_0 \in \mathcal{W}_0$, the error between the output y(t) and the reference signal $y_{ref}(t)$ satisfies $|y(t) - y_{ref}(t)| \to 0$."



- The problem is unsolvable for y_{ref} and w_d with large amplitudes, which corresponds to large $||v_0||!$
- No limitations for x_0 .

Solvability conditions and control law







Main Assumptions

$$\begin{split} \dot{x}(t) &= Ax(t) + B\phi(u(t)) + B_d w_d(t), \quad x(0) = x_0, \\ y(t) &= B^* x(t) \end{split}$$

Assumption 1

We assume

- The operator A generates a C₀-semigroup T(t) of contractions on X,
- $B \in \mathcal{L}(\mathbb{R}, X)$ and $B_d \in \mathcal{L}(\mathbb{R}^{n_d}, X)$,
- The operator A κBB* generates a strongly stable contraction semigroup T_{-κBB*}(t) for any κ > 0.

Theorem 2

Consider the compact set $\mathcal{W}_0 \subset \mathbb{R}^q$. Under the given Assumption 1, the semi-global output regulation problem is solvable if there exist $\Pi \in \mathcal{L}(\mathbb{R}^q, X)$ with $\mathcal{R}(\Pi) \subset D(A)$ and $\Gamma \in \mathbb{R}^{1 \times q}$ such that they solve the regulator equations

$$\Pi S = A\Pi + B\Gamma + B_d E$$
$$0 = B^*\Pi + F$$

and there exists a $\delta > 0$ such that $\sup_{t \ge 0} |\Gamma v(t)| \le 1 - \delta$ for all $v(t) = e^{St}v_0$ with $v_0 \in \mathcal{W}_0$. In this case, for any $\kappa > 0$ the feedback law

$$u(t) = -\kappa y(t) + (\kappa B^* \Pi + \Gamma) v(t)$$

solves the semi-global output regulation problem.

- Our main result is an infinite-dimensional generalization of a result of [Saberi et al., 2003].
- Our approach for showing the asymptotic convergence of the regulation error is motivated by the techniques in [Curtain and Zwart, 2020].

Outline of the Proof:

- By using strong stabilizability of the system by negative output feedback and uniform Lipschitz continuity of the saturation function, we have from [Curtain and Zwart, 2016] that the origin of the system $\dot{x}(t) = Ax(t) + B\phi(-\kappa y(t))$, $x(0) = x_0$ is globally asymptotically stable.
- We introduce a variable $\xi(t) = x(t) \Pi v(t)$ which is the mild solution of

$$\dot{\xi}(t) = A\xi(t) + B[\phi(-\kappa B^*\xi(t) + \Gamma v(t)) - \Gamma v(t)]$$

$$\xi(0) = \xi_0.$$

on X.

By using the assumptions of the theorem, we can show that $\xi(t)=x(t)-\Pi v(t)$ decays to zero asymptotically. This implies

$$y(t) - y_{ref}(t) = B^* x(t) - y_{ref}(t)$$
$$= \underbrace{B^* \xi(t)}_{\to 0} + \underbrace{(B^* \Pi + F) v(t)}_{= 0} \to 0$$

as $t \to \infty$ due to the fact that Π satisfies $B^*\Pi + F = 0$ by assumption.

Lemma 3 ([Byrnes et al., 2000])

Suppose Assumption 1 holds. The regulator equations are solvable if and only if $\pm i\omega_k$ do not coincide with the transmission zeros of the system (A, B, B^*) .

- Equivalently, $G(\pm i\omega_k) \neq 0$ for all k, where $G(\cdot) = B^*(\cdot I A)^{-1}B$ is the transfer function of the system (A, B, B^*) .
- In this case, Π and Γ have explicit formulas determined by

$$\Pi \Phi_k = (i\mu_k - A)^{-1} (B\Gamma + B_d E) \Phi_k$$

$$\Gamma \Phi_k = -G(i\mu_k)^{-1} (B^*(i\mu_k - A)^{-1} B_d E + F) \Phi_k,$$

 $k = 1, 2, \cdots, q$, where $i\mu_k$ and Φ_k are the eigenvalues and the corresponding orthonormal eigenvectors of S







We consider a dynamic model of a flexible satellite [Bontsema et al., 1988, He and Ge, 2015]

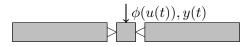


Figure: A flexible satellite model

$$\begin{split} \ddot{w}_{l}(\xi,t) + w_{l}^{\prime\prime\prime\prime}(\xi,t) + 5\dot{w}_{l}(\xi,t) &= 0, \ -1 < \xi < 0, t > 0, \\ \ddot{w}_{r}(\xi,t) + w_{r}^{\prime\prime\prime\prime}(\xi,t) + 5\dot{w}_{r}(\xi,t) &= 0, \ 0 < \xi < 1, t > 0, \\ \ddot{w}_{c}(t) &= w_{l}^{\prime\prime\prime}(0,t) - w_{r}^{\prime\prime\prime}(0,t) + \boldsymbol{\phi}(\boldsymbol{u}(t)) + w_{d}(t), \\ \ddot{\theta}_{c}(t) &= -w_{l}^{\prime\prime}(0,t) + w_{r}^{\prime\prime}(0,t). \end{split}$$

with measured output

$$y(t) = \dot{w}_c(t).$$

The boundary conditions are given by

$$w_l''(-1,t) = 0, \qquad w_l'''(-1,t) = 0,$$

$$w_r''(1,t) = 0, \qquad w_r'''(1,t) = 0,$$

$$\dot{w}_l(0,t) = \dot{w}_r(0,t) = \dot{w}_c(t),$$

$$\dot{w}_l'(0,t) = \dot{w}_r'(0,t) = \dot{\theta}_c(t).$$



• The satellite model can be written in the form [Govindaraj et al., 2023]

$$\dot{x}(t) = Ax(t) + B\phi(u(t)) + B_d w_d(t), \quad x(0) = x_0,$$

 $y(t) = B^*x(t)$

where the operator A generates an exponentially stable contraction semigroup on the state space $X = L^2(-1,0;\mathbb{R}^2) \times L^2(0,1;\mathbb{R}^2) \times \mathbb{R}^2.$

• It can be also verified that $A - \kappa BB^*$ generates an exponentially stable contraction semigroup on X for any $\kappa > 0$ [Govindaraj et al., 2020].



Our goal is to track the reference signal $y_{ref}(t) = 0.09 \sin (1.5t)$ and reject the disturbance $w_d(t) \equiv 0.08$. We choose the exosystem

$$\dot{v}(t) = \begin{bmatrix} 0 & 1.5 & 0 \\ -1.5 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} v(t), \quad v(0) = \begin{bmatrix} 0 \\ 0.09 \\ 0.08 \end{bmatrix},$$
$$y_{ref}(t) = -\begin{bmatrix} 1 & 0 & 0 \end{bmatrix} v(t),$$
$$w_d(t) = \begin{bmatrix} 0 & 0 & 1 \end{bmatrix} v(t).$$

- The eigenvalues are given by $\{0, \pm 1.5i\}$.
- Moreover, the satellite system does not have any transmission zeros at 0, 1.5i and -1.5i [Govindaraj et al., 2023]

The control parameters Γ and Π can be obtained by using

$$\Pi \Phi_k = (i\mu_k - A)^{-1} (B\Gamma + B_d E) \Phi_k$$

$$\Gamma \Phi_k = -G(i\mu_k)^{-1} (B^*(i\mu_k - A)^{-1} B_d E + F) \Phi_k,$$

and the formulas lead to

$$\begin{split} \Gamma v(t) &= 0.09 |G(1.5i)^{-1}| \sin(1.5t+\theta) + 0.08, \\ \theta &= \tan^{-1}(\beta/\alpha), \alpha = \mathsf{Re}(G(i\omega)), \beta = -\mathsf{Im}(G(i\omega)) \\ B^* \Pi v(t) &= 0.09 \sin(1.5t) \end{split}$$

in the control law

$$u(t) = -\kappa y(t) + (\kappa B^* \Pi + \Gamma) v(t)$$

where $G(\cdot)$ is the transfer function of the satellite system.

Output tracking is achieved using the control input $u(t) = -\kappa y(t) + (\kappa B^* \Pi + \Gamma) v(t).$

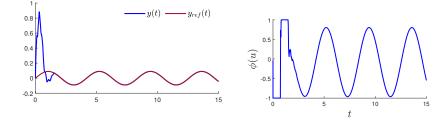


Figure: The output y(t) (blue), reference $y_{ref}(t)$ (red), and the saturated control input $\phi(u(t))$ (right) for $\kappa = 100$.

Conclusion

- We considered class of strongly stabilizable infinite-dimensional linear systems with collocated actuators and sensors subject to input saturation.
- We proposed a linear output feedback control law that solves the semi-global output regulation problem.
- The results are illustrated with an example of a flexible satellite model subject to input saturation.

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